

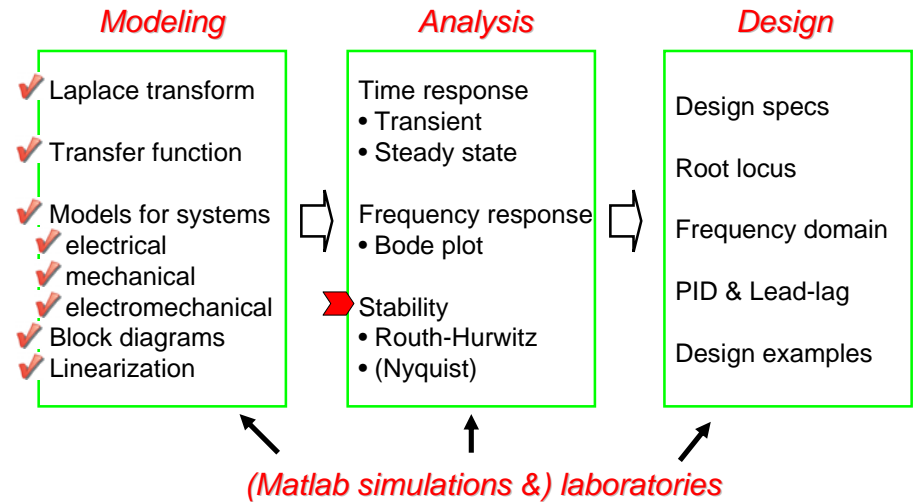
ME451: Control Systems

Lecture 9 Stability

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1

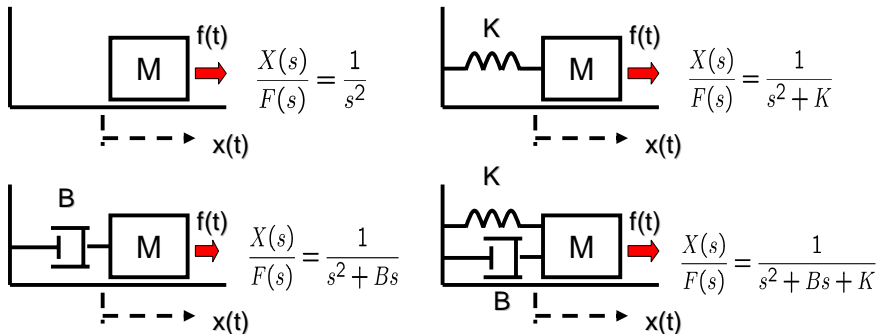
Course roadmap



2

Simple mechanical examples

- We want mass to stay at $x=0$, but wind gave some initial speed ($F(t)=0$). What will happen?



- How to characterize different behaviors with TF?

3

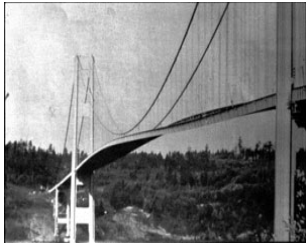
Stability

- Utmost important specification in control design!
- Unstable systems have to be stabilized by feedback.
- Unstable closed-loop systems are useless.
 - What happens if a system is unstable?
 - may hit mechanical/electrical “stops” (saturation)
 - may break down or burn out

4

What happens if a system is unstable?

Tacoma Narrows Bridge (July 1-Nov.7, 1940)



Wind-induced vibration



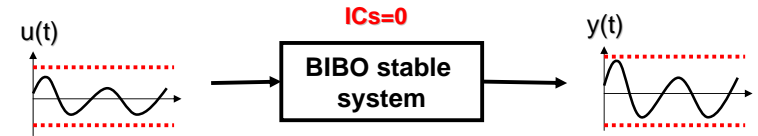
Collapsed!

2008...

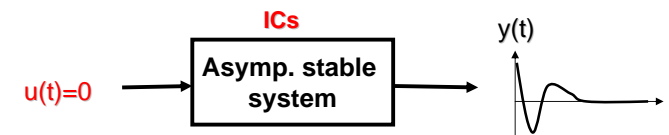


Mathematical definitions of stability

- **BIBO** (Bounded-Input-Bounded-Output) **stability** :
Any bounded input generates a bounded output.



- **Asymptotic stability** :
Any ICs generates $y(t)$ converging to zero.



Some terminologies

$$G(s) = \frac{n(s)}{d(s)}$$

Ex. $G(s) = \frac{(s-1)(s+1)}{(s+2)(s^2+1)}$

- **Zero** : roots of $n(s)$ (Zeros of G) = ± 1
- **Pole** : roots of $d(s)$ (Poles of G) = $-2, \pm j$
- **Characteristic polynomial** : $d(s)$
- **Characteristic equation** : $d(s)=0$

Stability condition in s-domain (Proof omitted, and not required)

For a system represented by a transfer function $G(s)$,

system is **BIBO stable**



All the poles of $G(s)$ are in the open left half of the complex plane.



system is **asymptotically stable**

"Idea" of stability condition

Example $y'(t) + \alpha y(t) = u(t), y(0) = y_0$

➔ $sY(s) - y(0) + \alpha Y(s) = U(s)$

➔ $Y(s) = \frac{1}{s + \alpha}(U(s) + y(0))$

Asym. Stability: $y(t) = \mathcal{L}^{-1}\{Y(s)\} = \mathcal{L}^{-1}\left\{\frac{1}{s + \alpha}y(0)\right\} = e^{-\alpha t}y(0) \rightarrow 0 \Leftrightarrow \text{Re}(\alpha) > 0$
(U(s)=0)

BIBO Stability: $y(t) = \mathcal{L}^{-1}\{Y(s)\} = \mathcal{L}^{-1}\{G(s)U(s)\} = \int_0^t y(\tau)u(t-\tau)d\tau = \int_0^t e^{-\alpha\tau}u(t-\tau)d\tau$
(y(0)=0)

$$|y(t)| \leq \int_0^t |e^{-\alpha\tau}| |u(t-\tau)| d\tau \leq \int_0^t |e^{-\alpha\tau}| d\tau \cdot u_{max}$$

Bounded if $\text{Re}(\alpha) > 0$

Second order impulse response- Underdamped and Undamped

$$\delta(t) \rightarrow \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \rightarrow h(t)$$

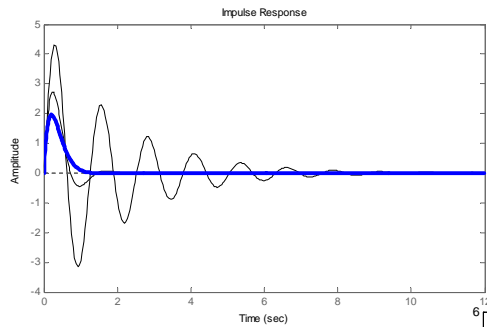
$$p_{1,2} = -\zeta\omega_n \pm j\omega_n\sqrt{1 - \zeta^2}$$

- $\zeta > 1$... Overdamped
- $\zeta = 1$... Critically damped
- $1 > \zeta > 0$... Underdamped
- $\zeta = 0$... Undamped

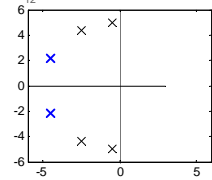
$$h(t) = \frac{\omega_n}{\sqrt{1 - \zeta^2}} e^{-\zeta\omega_n t} \sin(\omega_n\sqrt{1 - \zeta^2}t) 1(t)$$

Second order impulse response – Underdamped and Undamped

Changing ζ / Fixed ω_n

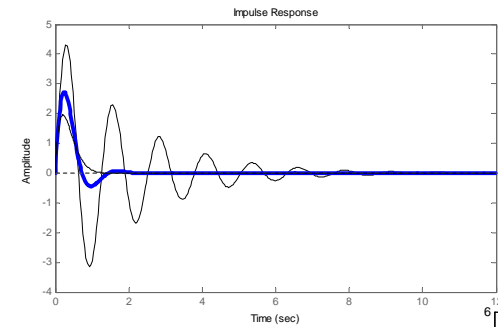


$$h(t) = \frac{\omega_n}{\sqrt{1 - \zeta^2}} e^{-\zeta\omega_n t} \sin(\omega_n\sqrt{1 - \zeta^2}t) 1(t)$$

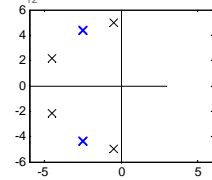


Second order impulse response – Underdamped and Undamped

Changing ζ / Fixed ω_n

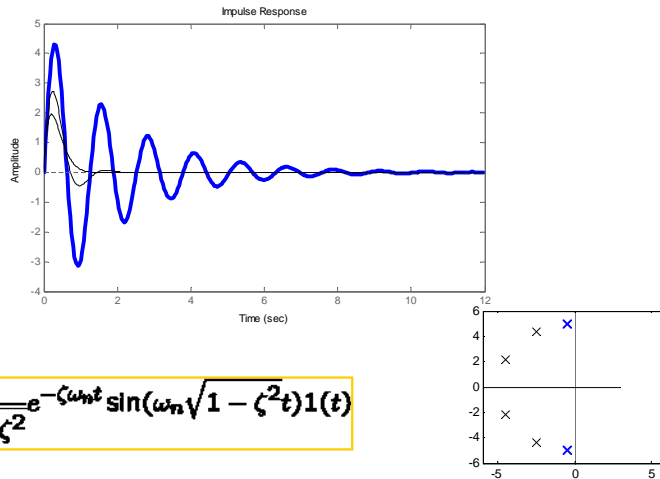


$$h(t) = \frac{\omega_n}{\sqrt{1 - \zeta^2}} e^{-\zeta\omega_n t} \sin(\omega_n\sqrt{1 - \zeta^2}t) 1(t)$$



Second order impulse response –

Underdamped and Undamped
Changing ζ / Fixed ω_n



$$h(t) = \frac{\omega_n}{\sqrt{1-\zeta^2}} e^{-\zeta\omega_n t} \sin(\omega_n \sqrt{1-\zeta^2} t) 1(t)$$

Remarks on stability

- For a general system (nonlinear etc.), BIBO stability condition and asymptotic stability condition are different.
- For **linear time-invariant (LTI) systems** (to which we can use Laplace transform and we can obtain a transfer function), the conditions happen to be the same.
- In this course, we are interested in only LTI systems, we use simply “**stable**” to mean both BIBO and asymptotic stability.

14

Remarks on stability (cont'd)

- Marginally stable** if
 - $G(s)$ has no pole in the open RHP (Right Half Plane), &
 - $G(s)$ has at least one simple pole on $j\omega$ -axis, &
 - $G(s)$ has no multiple poles on $j\omega$ -axis.

$$G(s) = \frac{1}{s(s^2 + 4)(s + 1)}$$

Marginally stable

$$G(s) = \frac{1}{s(s^2 + 4)^2(s + 1)}$$

NOT marginally stable

- Unstable** if a system is neither stable nor marginally stable.

15

Examples

- Repeated poles

$$\mathcal{L}^{-1} \left[\frac{2\omega s}{(s^2 + \omega^2)^2} \right] = t \sin \omega t \quad \mathcal{L}^{-1} \left[\frac{s^2 - \omega^2}{(s^2 + \omega^2)^2} \right] = t \cos \omega t$$

$$\dots = t^2 \sin \omega t \quad \dots = t^2 \cos \omega t$$

- Does marginal stability imply BIBO stability?

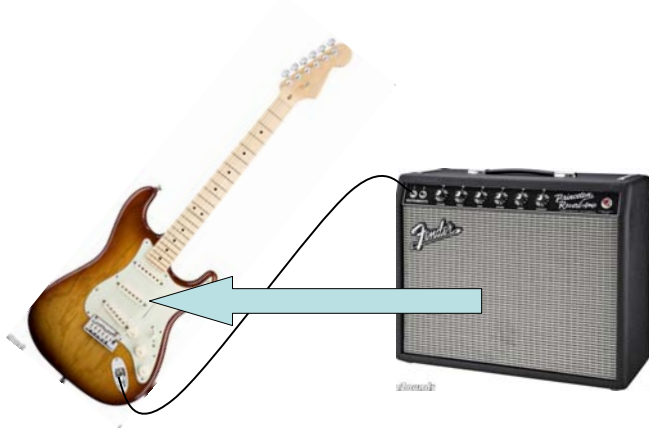
- TF: $G(s) = \frac{2s}{(s^2 + 1)}$

- Pick $u(t) = \sin t \xrightarrow{\mathcal{L}} U(s) = \frac{1}{(s^2 + 1)}$

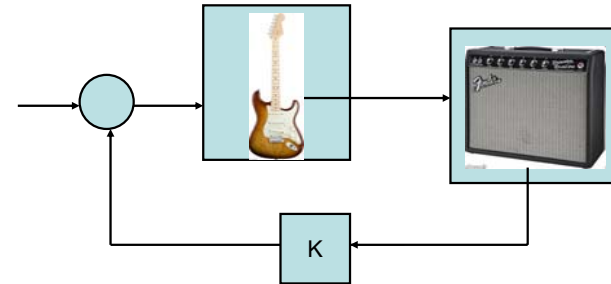
- Output $\mathcal{L}^{-1} \left[Y(s) = G(s)U(s) = \frac{2s}{(s^2 + 1)^2} \right] = t \sin t$

16

Feedback Technique



Positive Feedback

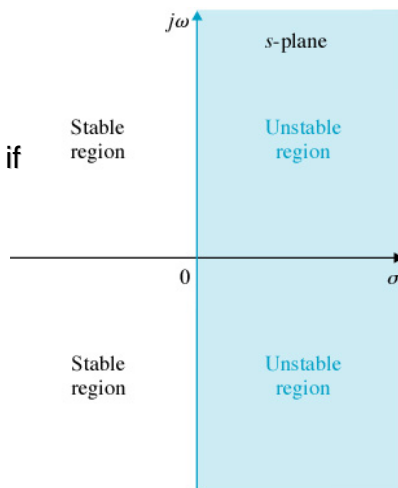


K will depend on the distance between the guitar and the amplifier.

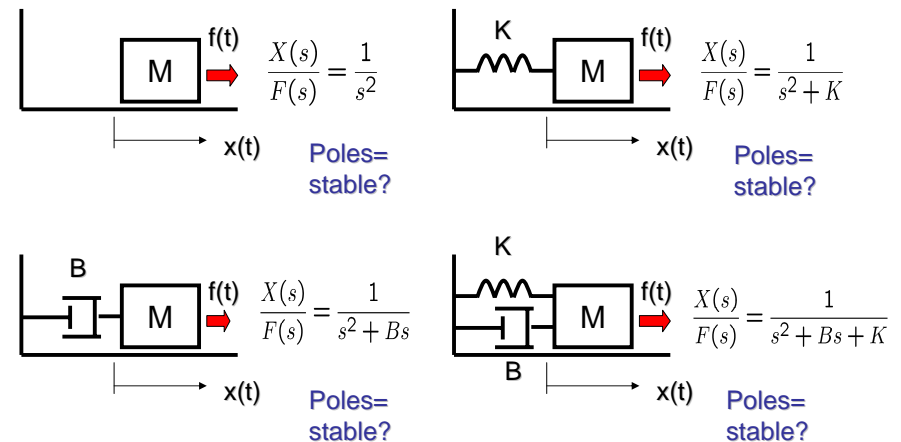
Stability summary

Let s_i be poles of G .
Then, G is ...

- **(BIBO, asymptotically) stable** if $\text{Re}(s_i) < 0$ for all i .
- **marginally stable** if
 - $\text{Re}(s_i) \leq 0$ for all i , and
 - simple root for $\text{Re}(s_i) = 0$
- **unstable** if it is neither stable nor marginally stable.



Mechanical examples: revisited



Examples

$G(s)$	Stable/marginally stable /unstable
$\frac{5(s+2)}{(s+1)(s^2+s+1)}$?
$\frac{5(-s+2)}{(s+1)(s^2+s+1)}$?
$\frac{5}{(s-2)(s^2+3)}$?
$\frac{s^2+3}{(s+1)(s^2-s+1)}$?
$\frac{1}{(s+1)(s^2+1)^2}$?
$\frac{1}{(s^2-1)(s+1)}$???

21

Summary and Exercises

- Stability for LTI systems
 - (BIBO and asymptotically) stable, marginally stable, unstable
 - Stability for $G(s)$ is determined by poles of G .
- Next
 - **Routh-Hurwitz stability criterion** to determine stability without explicitly computing the poles of a system.
- Exercises
 - Solve examples in the previous slide.

22